

## Dual Fractional Power Approximation for Nonlinear Functions

Yasuhiro KOBAYASHI\*, Masaaki OHKITA\*  
and Michio INOUE\*

### Abstract

For accurate analog simulation of continuous functions, it is advisable to use approximations of the form:

$$f(x) \approx \alpha_0 + \alpha_1 x + \alpha_2 x^{\beta_1} + \alpha_3 x^{\beta_2}$$

in a full range of an input variable. The exponents  $\beta_1$ ,  $\beta_2$  are given by real numbers according to the nature of  $f(x)$ . The accuracy is comparable to that in approximation using Chebyshev polynomials up to the fifth-degree. Especially, for odd functions, this approximation can achieve good accuracy over an extended interval.

### 1. Introduction

In the usual analog simulation, uniformly continuous functions have been divided into several subintervals wherein the functions are approximated by linear segments [1], quadratic arcs[2], or their combinations[3]. Unfortunately, these systems yield considerable errors near the endpoints and between them.

An approach is to use uniform(or non-breakpoint) approximation of continuous functions over the full range of an input variable. With this policy, an economical mathematical system, termed the *single fractional power approximation*, was introduced [4],[5]. To improve that accuracy or to extend that variable-range, in this paper, a fractional power is added to that system. This system, termed the *dual fractional power approximation*, can improve that accuracy by about a hundred times. Since a fractional power term behaves like an ordinary power series alternating in sign, one can expect better accuracy than Chebyshev approximations for odd functions. Parameter values must be given so as to satisfy a minimax error condition, although it cannot be achieved directly. We must correct their approximate values by iteration. The starting values can be determined by a method of undetermined coefficients or that of

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\* Department of Electrical Engineering, Faculty of Engineering, Tottori University

interpolation. For convenience of optimization, the parameter-determination begins with interpolation of a given function at geometrically arranged points in a specified interval. Details are as follows;

2. Parameter-determination and its optimization

Consider that a single-valued continuous function is approximated by  $f(x) = \alpha_0 + \alpha_1 x + \alpha_2 x^{\beta_1} + \alpha_3 x^{\beta_2}$  in the interval  $[0, 1]$ . An optimal condition is made so that the approximation satisfies null error at both endpoints and minimizes the maximum magnitude of deviations within this interval. An actual criterion is given so that the error curve possesses an equal magnitude oscillatory behavior with proper alternation in this interval. Since the minimax error magnitude  $\mu^*$  and the true extreme locations cannot be evaluated straightforwardly, the minimax condition in a Chebyshev sense is modified here.

For convenience of interpolation, suppose a set of distinct points  $X_i$  ( $i = 1, 2, 3, 4$ ) satisfying a geometrical relation  $X_i = \lambda^{(i-5)}$  with  $\lambda > 1$  on the x-axis within the interval (See Fig. 1). If the optimal approximation yields certain deviations  $\rho_i$  ( $i = 1, 2, 3, 4$ ) at these points, the following simultaneous equation is established.

$$\left. \begin{aligned} f(0) &= \alpha_0, \\ f(X_1) + \rho_1 &= \alpha_0 + \alpha_1 \lambda^{(i-5)} + \alpha_2 \lambda^{(i-5)\beta_1} + \alpha_3 \lambda^{(i-5)\beta_2}, \\ f(1) &= \alpha_0 + \alpha_1 + \alpha_2 + \alpha_3. \end{aligned} \right\} (1)$$

With the transformation:  $P = \lambda^{(1-\beta_1)}$ ,  $Q = \lambda^{(1-\beta_2)}$ , (2) one can rearrange Eq. (1) so that

$$F_i = \alpha_2 [P^{(5-i)} - 1] + \alpha_3 [Q^{(5-i)} - 1], \quad i = 1, 2, 3, 4, \quad (3)$$

where  $F_i = \lambda^{(5-i)} [f(X_i) - f(0) + \rho_i] - [f(1) - f(0)]$ .

Moreover, upon putting

$$H_1 = F_1 - 4F_2 + 6F_3 - 4F_4, \quad H_2 = F_2 - 3F_3 + 3F_4, \quad H_3 = F_3 - 2F_4, \quad H_4 = F_4, \quad (4)$$

one can rewrite Eq. (3) as  $H_i = \alpha_2 (P-1)^{(5-i)} + \alpha_3 (Q-1)^{(5-i)}$ . (5)

Equation (5) is then modified so that

$$H_1 - [(P-1) + (Q-1)]H_2 + (P-1)(Q-1)H_3 = 0, \quad H_2 - [(P-1) + (Q-1)]H_3 + (P-1)(Q-1)H_4 = 0. \quad (6)$$

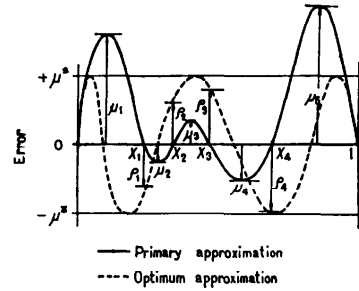


Fig. 1 Diagram of the interpolated approximations.

Hence  $(P-1)+(Q-1) = \frac{H_2 H_3 - H_1 H_4}{H_3^2 - H_2 H_4}$ ,  $(P-1)(Q-1) = \frac{H_2^2 - H_1 H_3}{H_3^2 - H_2 H_4}$ ,  $H_3^2 - H_2 H_4 \neq 0$ . (7)

This means that the  $(P-1)$  value, corresponding to the fractional exponent  $\beta_1$ , can be evaluated as either of the two roots of a quadratic equation. Therefore,

$P-1 = (M + \sqrt{M^2 - 4N})/2$ ,  $Q-1 = (M - \sqrt{M^2 - 4N})/2$ ,  $M^2 - 4N > 0$ , (8)

where  $M = \frac{H_2 H_3 - H_1 H_4}{H_3^2 - H_2 H_4}$ ,  $N = \frac{H_2^2 - H_1 H_3}{H_3^2 - H_2 H_4}$ ,  $H_3^2 - H_2 H_4 \neq 0$ .

Then, the parameter values can be evaluated as

$\alpha_0 = f(0)$ ,  $\alpha_1 = f(1) - f(0) + (H_3 - H_4 M)/N$ ,  
 $\alpha_2 = \frac{2H_3 - (M - \sqrt{M^2 - 4N})H_4}{(M + \sqrt{M^2 - 4N})\sqrt{M^2 - 4N}}$ ,  $\alpha_3 = \frac{-2H_3 + (M + \sqrt{M^2 - 4N})H_4}{(M - \sqrt{M^2 - 4N})\sqrt{M^2 - 4N}}$ , (9)  
 $\beta_1 = -\log[(2 + M + \sqrt{M^2 - 4N})/2\lambda]/\log \lambda$ ,  $\beta_2 = -\log[(2 + M - \sqrt{M^2 - 4N})/2\lambda]/\log \lambda$ .

Note that these parameter values (except for  $\alpha_0$ ) are given as functions of  $\rho_i$  values, though difficult to evaluate analytically. For convenience, the  $\rho_i$  values are estimated in terms of neighboring extreme deviations  $\mu_i, \mu_{i+1}$ . At the first step of computation, all  $\rho_i$  values are given as zero; and at the second and thereafter, they are estimated so that  $\rho_i = W(\mu_i + \mu_{i+1})$  with a weight  $W$ . The  $\rho_i$  values are accumulated with each step. By substituting such values into Eq. (1) iteratively, the parameter values can be modified. The modification must be repeated until all extreme deviations are leveled. Fortunately, the computation converges without oscillation and produces the optimal parameters. An experimental verification is given later.

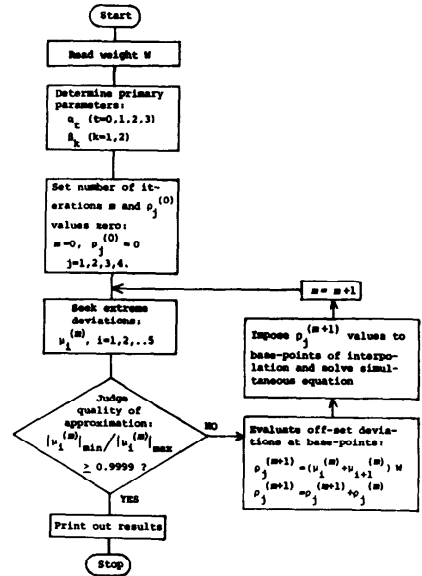


Fig. 2 Flowchart for the minimax dual fractional power approximation.



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59 17 CONTINUE
60 DEB=(EEMIN/EEMAX)
61 DO 18 I=1,4
62 18 FFL(I)=(EY(I)+EY(I+1))/U
63 RETURN
64 END

** MINIMAX DUAL FRACTIONAL POWER APPROXIMATION **

FUNCTION=COSH(X) WEIGHT= 0.3000000E 00

1 SUBROUTINE OUT
2 DOUBLE PRECISION A(6),XX(5),EY(5),U
3 COMMON /AA/N,AM,V,LP
4 COMMON /BB/DEB
5 COMMON /CC/A
6 COMMON /DD/XX,EY
7 WRITE(6,1) LP,DEB
8 WRITE(6,2) A(1),A(2)
9 WRITE(6,3) A(3),A(4)
10 WRITE(6,4) A(5),A(6)
11 WRITE(6,5) (JI,XX(JI),EY(JI),JI=1,5)
12 FORMAT(/'/7,'ITERATIVE CORRECTION=',I3.
13 1 T36,'DEB',E15.7)
14 2 FORMAT(1H ,5X,3HA0=,D20.11,5X,3HA1=,D20.11)
15 3 FORMAT(1H ,5X,3HA2=,D20.11,5X,3HA3=,D20.11)
16 4 FORMAT(1H ,4X,4R P =,D20.11,4X,4R Q =,D20.11)
17 5 FORMAT(1H ,5X,3H X(,I2,2H)=,O17.10,5X,
18 SHEMAX=,D17.10)
19 RETURN
20 END

ITERATIVE CORRECTION= 0 DEB= 0.2630651E-01
A0= 0.1000000000E 01 A1= 0.39815809449E-03
A2= 0.5021024190E 00 A3= 0.40380057686E-01
P = 0.2004553890E 01 Q = 0.4142957495E 01
X( 1)= 0.6323852539E-01 EMAX=-0.827515137E-05
X( 2)= 0.3190979004E 00 EMAX= 0.2176577406E-06
X( 3)= 0.4964754301E 00 EMAX=-0.2540126625E-06
X( 4)= 0.6469665527E 00 EMAX= 0.7825217406E-06
X( 5)= 0.9165588379E 00 EMAX=-0.6557847031E-05

ITERATIVE CORRECTION= 29 DEB= 0.9999077E 00
A0= 0.1000000000E 01 A1= 0.2105676319E-03
A2= 0.5020274414E 00 A3= 0.4084262577E-01
P = 0.20034553277E 01 Q = 0.41399901639E 01
X( 1)= 0.3708618164E-01 EMAX=-0.2751953695E-05
X( 2)= 0.2061242676E 00 EMAX= 0.2752206814E-05
X( 3)= 0.4655187988E 00 EMAX=-0.2752131198E-05
X( 4)= 0.7389050293E 00 EMAX= 0.2752054193E-05
X( 5)= 0.9436755371E 00 EMAX=-0.2752207854E-05

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Fig. 3 Program for the minimax dual fractional power approximation and its results.

3. Numerical experimentation

A computing procedure is outlined in Fig. 2 and detailed in Fig. 3. This program is provided for a TOSBAC 3400 model 21 digital computer. An investigation of a ratio of the minimum to the maximum absolute deviations shows how the optimization is achieved. Some of the resulting approximations are enumerated in Table 1 with each maximum absolute deviation.

Table 1 Minimax dual fractional power approximations furnishing more than 0.9999 of error-balance\* in the interval [0, 1].

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Functions	Minimax Dual Fractional Power Approximations	Max. Abs. Errors	Full Scale Errors(%)
exp x	$1.000000000 + 1.002261347 x + 0.571024656 x^2 + 0.144895825 x^3 + 0.694510965 x^4$	$0.2193 \times 10^{-4}$	$0.8068 \times 10^{-3}$
sinh x	$0.000000000 + 1.000004326 x + 0.167050480 x^3 + 0.008146388 x^5 + 0.111804903 x^7$	$0.1839 \times 10^{-6}$	$0.1565 \times 10^{-4}$
cosh x	$1.000000000 + 0.000210568 x + 0.502027441 x^2 + 0.03455328 x^4 + 0.040842626 x^6 + 0.139990164 x^8$	$0.2752 \times 10^{-5}$	$0.1783 \times 10^{-3}$
$\sin \frac{x}{2}$	$0.000000000 + 1.570689531 x - 0.658293028 x^3 - 0.12202756 x^5 + 0.087603497 x^7 + 0.708008110 x^9$	$0.4318 \times 10^{-5}$	$0.4318 \times 10^{-3}$
$\cos \frac{x}{2}$	$1.000000000 - 0.003533570 x - 1.281127771 x^2 - 0.027488275 x^4 + 0.284661341 x^6 + 0.624865310 x^8$	$0.4250 \times 10^{-4}$	$0.4250 \times 10^{-2}$
erf x	$0.000000000 + 1.127857706 x - 0.660091651 x^3 - 0.214527565 x^5 + 0.374934738 x^7 + 0.760594092 x^9$	$0.1871 \times 10^{-4}$	$0.2220 \times 10^{-2}$

\* Error-balance is defined by the ratio of the minimum to the maximum in absolute extreme deviations.

4. Discussion

(1) Characterization of the fractional power system

Since each fractional power term has two parameters, the use of dual fractional powers can achieve a degree of accuracy in Chebyshev approximations using the fifth-degree polynomials. If a given function behaves like an alternating series, this

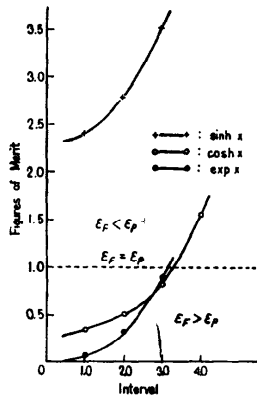


Fig. 4  
 Figures of Merit for both approximations using dual fractional powers and ordinary powers in the same interval. (Figure of Merit is defined by  $\epsilon_P / \epsilon_F$ ; where  $\epsilon_P$  and  $\epsilon_F$  show the absolute errors due to the minimax polynomial approximation of the fifth-degree and due to the dual fractional power approximation, respectively.)

mathematical system can achieve better approximation than ordinary power series. One of the advantageous features results from the capability of extended variable-range applications. This permits wide selection of variable-scaling in analog computers. The usefulness of the fractional power approximations can be found in Fig. 4.

(ii) On the local minimax approximation

The resulting approximation exhibits the equal-magnitude oscillatory behavior for its error. Then, the error satisfies six distinct roots in the closed interval  $[a, b]$ , and attains an extreme deviation between two roots. The first and last extreme deviations in a Chebyshev approximation are forced out. This condition exhibits a local minimax approximation between

the first and last null error points of the Chebyshev approximation. Principally, the maximum error magnitude is somewhat larger than that of the minimax approximation. In the diagram of Fig. 5, the characterization of the local minimax approximation is compared with that of the Chebyshev sense approximation.

(iii) On the optimization approach

The usual techniques have required repetition of tedious numerical calculations. This is because increments of the parameter values have been evaluated as a set of solutions of a linear problem between the shift of the extremum points and the deviation of the extremum magnitude. Therefore, all matrix-elements are necessarily refreshed in every cycle of correction. For economical calculation, an improved method

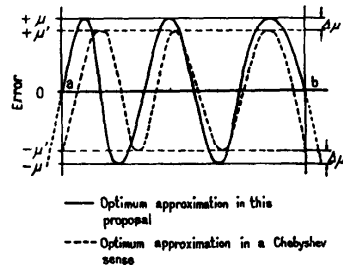


Fig. 5 Typical error-curves of both approximations in the same interval  $[a, b]$ .

is proposed, which repeats interpolation at the base-points with off-set deviations. Proper deviations are given in terms of the weighted algebraic sum of neighboring extreme deviations. With the pertinent weight (a number  $1/3 \sim 1/20$  is available), we can achieve rapid convergence of the computation. Figure 6 shows how the approximation to the function  $\cosh x$  ( $0 \leq x \leq 1$ ) is optimized with such weights. In this process, the off-set deviations  $\rho_i$  at the base-points converge to the proper values as shown in Fig. 7. Obviously, the convergence of the quality of approximation corresponds to that of the  $\rho_i$  values. This fact shows that there is a practical reason in our assumption that  $\rho_i = W(\mu_i + \mu_{i+1})$ . In some applications, this algorithm fails sometimes in the optimization when there is no true extreme deviation between two preliminary arranged base-points. Arrangement of the base-points with equal spaces will succeed in such a case, although the parameter values cannot be evaluated explicitly.

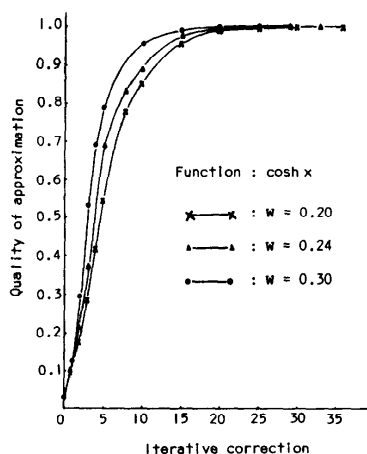


Fig. 6 Convergence of the approximation to the function  $\cosh x$  ( $0 \leq x \leq 1$ ) for the weights 0.20, 0.24 and 0.30.

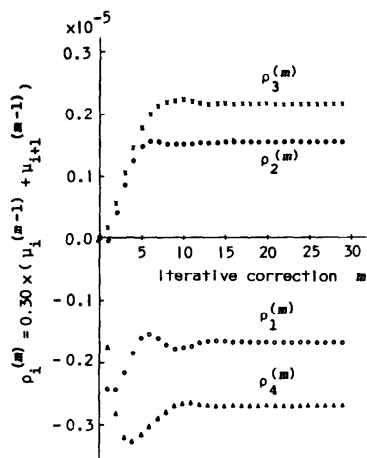


Fig. 7 Convergence of the off-set deviations with the weight 0.30 for the function  $\cosh x$  ( $0 \leq x \leq 1$ ).

## 5. Conclusions

The usefulness of the dual fractional power approximation has been clarified. Since each of the fractional power terms contains two parameters, one can economize on two ordinary power terms without diminishing the accuracy. Therefore, the degree of accuracy becomes comparable to that in approximation using the Chebyshev-polynomials up to the fifth-degree. For odd functions, especially, this approximation

can achieve good accuracy over an extended interval. An algorithm of the optimal parameter-determination can be completed, which achieves a minimax condition with null error terminals.

This approximation system will serve as an economical and accurate simulation in analog computer applications.

#### References

- [1] G. A. Korn and T. M. Korn, *Electronic Analog and Hybrid Computers*, New York, McGraw-Hill, 1964, pp. 233 ~ 240.
- [2] Y. Kobayashi, M. Ohkita and T. Chiba, "Arbitrary Function Generator," Japan. Pat. no. 9537, 1975.
- [3] A. Nathan, "Linear and Nonlinear Interpolators," *IEEE Trans. Electron. Comput.*, vol. EC-12, pp. 526 ~ 532, 1963.
- [4] Y. Kobayashi, M. Ohkita and M. Inoue, "Fractional Power Approximation and Its Generation," *Trans. IMACS*, vol. 18, no. 12, pp. 115 ~ 122, 1976.
- [5] Y. Kobayashi, M. Ohkita and M. Inoue, "Generation of Functions Approximated by a Fractional Power," *Trans. IEE, Japan*, vol. 96, no. 8, pp. 173 ~ 178, 1976.